LAPLACE TIME TUNING TRANSFORM IN DIGITAL SIGNAL PROCESSING

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Abstract: In this paper, we derive a series of new theorems and formulae on certain type of multi-series by equating closed and summation solutions of generalized difference equation with trignometric functions. We generate Discrete Laplace Transform with Time Tuning Factor ℓ in the field of Digital Signal Processing (DSP). Suitable examples with relevant diagrams which are generated and verified using MATLAB are inserted to validate our main results.

Key words: Generalized difference equation, Summation solution, Closed form solution, m-series, Laplace Transform and Time tuning factor.

AMS Subject classification: 39A70, 47B39, 39A10, 44A10, 65Q10, 92C55.

1. INTRODUCTION

Digital Signal Processing (DSP) has revolutionized many areas in science and engineering such as space, medicine, commerce, military, technology and communication. DSP is made effectively possible by Laplace Transform (LT) and Discrete Laplace Transform (DLT) which changes a signal in the time domain into frequency s-domain [15]. The Laplace transforms are practical in the view of fast decay factor e^{-sx} . Construction of polynomial filters for detection of peaks in periodic signals in DSP is developed in [16]. With numerical computation and MATLAB obtaining exact solutions for Dirichlet-Neumann inverse problem are discussed in [4]. In practice, many applications of Laplace Transform (LT) and Discrete Laplace Transform (DLT) are discussed by several authors [3, 9, 14, 10].

The LT and DLT are respectively defined as $L[f(t)] = \int_{0}^{\infty} f(t)e^{-st}dt$ and $L[f(n)] = \sum_{n=0}^{\infty} f(n)e^{-sn}$, s > 0. From the basic difference identity $\Delta^{-1}x_n \mid_{0}^{\infty} = \sum_{n=0}^{\infty} x_n$ [1], the DLT can be

expressed as $L[f(n)] = \Delta^{-1} f(n) e^{-sn}$. Let u(k) be input signal(functions) and ℓ be the time interval between two successive signals.

In [8], authors have defined a new type Laplace Transform as

$$L_{\ell}u(k) = \ell \Delta_{\ell}^{-1}u(k)e^{-sk} \mid_{0}^{\infty} = \ell \sum_{r=0}^{\infty} u(r\ell)e^{-sr\ell}.$$
 (1)

This transform is called as Generalized Laplace Transform (GLT) and it lies in between DLT and LT. The GLT becomes DLT and LT when $\ell = 1$ and $\ell \rightarrow 0$ respectively [2, 3]. If we take ℓ as time between two successive signals in DSP (1) becomes Laplace Time Tuning Transform (LTTT). To develop certain theories on LTTT we need to reveal the basic theory of Δ_ℓ and their inverse [5, 12].

The main definition of fractional difference equation (as done in [13]) is the V fractional sum of f(t) by

$$\Delta^{-\nu} f(t) = \frac{1}{\Gamma_{(\nu)}} \sum_{s=a}^{t-\nu} \frac{\Gamma_{(t-s)}}{\Gamma_{(t-s-(\nu-1))}} f(s), \text{ where } \nu > 0$$

When v = m is a positive integer, if we replace f(t) by u(k) and Δ by Δ_{ℓ} .

(i.e. $\Delta_{\ell} u(k) = u(k+\ell) - u(k), k \in [0,\infty), \ell > 0$), we arrive an m-series

$$\Delta_{\ell}^{-m}u(k) = \sum_{r=m}^{\left[\frac{k}{\ell}\right]} \frac{(r-1)^{(m-1)}}{(m-1)!} u(k-r\ell),$$
(2)

where $(r-1)^{(m-1)} = (r-1)(r-2)\cdots(r-m+1)$ and $\left[\frac{k}{\ell}\right]$ is the integer part of $\frac{k}{\ell}$. This m-series is a

numerical solution of the mth order generalized difference equation

$$\Delta_{\ell}^{m} v(k) = u(k), \ k \in [0, \infty), \ \ell > 0.$$
(3)

When m = 1, $k \to \infty$, replacing u(k) by $\ell u(k)e^{-sk}$ and by rearranging the terms, the equation (2) for $\ell u(k)e^{-sk}$ becomes LTTT given in (1).

Let $\ell > 0$ be time tuning factor of two successive signals and u(k) be a real valued function on $[0,\infty)$ representing input signal of a system in DSP. In this paper, by deriving several formulas on m-series to circular functions with respect to ℓ , we analyze the Laplace Time Tuning Transform in the field of Digital Signal Processing.

2. PRELIMINARIES

Before stating and proving our results, we present some notations, basic definitions and preliminary results which will be useful for further subsequent discussions. $\Delta_{\ell}^{-m}u(k) \parallel_{(m-1)\ell+j}^{k} = \Delta_{\ell}^{-1}\left(\cdots \Delta_{\ell}^{-1}u(k) \parallel_{j}^{k}\right) \parallel_{\ell+j}^{k} \cdots \parallel_{(m-1)\ell+j}^{k}$, where $\Delta_{\ell}^{-1}u(k) \parallel_{j}^{k} = u_{1}(k) = \Delta_{\ell}^{-1}u(k) - \Delta_{\ell}^{-1}u(j)$, $\Delta_{\ell}^{-1}\left(\Delta_{\ell}^{-1}u(k) \parallel_{j}^{k}\right) \parallel_{\ell+j}^{k} = u_{2}(k) = \Delta_{\ell}^{-1}u_{1}(k) - \Delta_{\ell}^{-1}u_{1}(\ell+j)$, and so on, $j = k - \left[\frac{k}{\ell}\right]\ell$, $\mathbb{N}_{\ell}(j) = \{j, \ell+j, 2\ell+j, \cdots\}$ and $\mathbb{N}_{1}(j) = \mathbb{N}(j)$. c_{j} is a constant for all $k \in \mathbb{N}_{\ell}(j)$ and for any positive integer m.

Also, $L_{m-1} = \{1, 2, ..., m-1\}$, $0(L_{m-1}) = \{\phi\}$, ϕ is empty set, $t(L_{m-1}) =$ set of all subsets of size t from the set L_{m-1} such that if $\{m_1, m_2, ..., m_t\} \in t(L_{m-1})$ then $m_1 < m_2 < ... < m_t$ for t = 1, 2, ..., m-1, $\wp(L_{m-1}) = \bigcup_{t=0}^{m-1} t(L_{m-1})$ is the power set of L_{m-1} , $\sum_{t=1}^{m-1} f(t) = 0$ for $m \le 1$, and $\prod_{i=2}^{t} f(i) = 1$ for $t \le 1$, and $\{m_t\} \in t(L_{m-1})$ means that $\{m_1, m_2, ..., m_t\} \in t(L_{m-1})$.

In [7] the authors have introduced generalized polynomial factorial $k_{\ell}^{(m)} = k(k-\ell)(k-2\ell)\cdots(k-(m-1)\ell)$. Using Stirling numbers of first kind s_r^m and second kind S_r^m , the following identities have been obtained:

$$(i) k_{\ell}^{(m)} = \sum_{r=1}^{m} s_{r}^{m} \ell^{m-r} k^{r}, (ii) k^{m} = \sum_{r=1}^{m} S_{r}^{m} \ell^{m-r} k_{\ell}^{(r)}, (iii) \Delta_{\ell} k_{\ell}^{(m)} = (m\ell) k_{\ell}^{(m-1)}.$$
(4)

Definition 2.1 [11] Let u(k), $k \in [0, \infty)$ be a real valued function. The generalized difference operator Δ_{ℓ} on u(k) is defined as;

$$\Delta_{\ell} u(k) = u(k+\ell) - u(k), \, k \in [0,\infty), \, \ell > 0, \tag{5}$$

and the inverse of Δ_{ℓ} on u(k) is defined as,

if
$$\Delta_{\ell} v(k) = u(k)$$
, then $v(k) = \Delta_{\ell}^{-1} u(k) + c_{j}$. (6)

Ingeneral, $\Delta_{\ell}^{-\nu} = \Delta_{\ell}^{-1} \left(\Delta_{\ell}^{-(\nu-1)} \right).$

Lemma 2.2 [7] Let p and q be any real numbers. If $1 - \cos p\ell \neq 0$, then

$$\Delta_{\ell}^{-1} \sin pk = \frac{\sin p(k-\ell) - \sin pk}{2(1 - \cos p\ell)} + c_{j}$$
(7)

and

$$\Delta_{\ell}^{-1} \cos pk = \frac{\cos p(k-\ell) - \cos pk}{2(1 - \cos p\ell)} + c_j \tag{8}$$

are solutions of (3) for $u(k) = \sin pk$ and $\cos pk$ respectively when m = 1.

Remark 2.3 Throughout this paper, we denote $P = p(n_2 - 2s_2) + q(n_3 - 2s_3)$ and

 $\overline{P} = p(n_2 - 2s_2) - q(n_3 - 2s_3)$. P and \overline{P} are depending on n_2 , n_3 , s_2 , s_3 , p and q.

To evaluate $\Delta_{\ell}^{-1}u(k)e^{-sk}$ (LTTT) we present the following lemmas and theorem.

Lemma 2.4 [6] Let v(k) and w(k) be two real valued functions. Then,

$$\Delta_{\ell}^{-1}[v(k)w(k)] = v(k)\Delta_{\ell}^{-1}w(k) - \Delta_{\ell}^{-1}[\Delta_{\ell}^{-1}w(k+\ell)\Delta_{\ell}v(k)].$$
(9)

Lemma 2.5 [11] Let $k \in [\ell, \infty)$. Then,

$$\Delta_{\ell}^{-1} e^{-k} \mid_{j}^{k} = \frac{e^{-k}}{e^{-\ell} - 1} - \frac{e^{-j}}{e^{-\ell} - 1}.$$
(10)

Lemma 2.6 [11] Let n be any non-negative integer. Then,

$$\Delta_{\ell}^{-1} k_{\ell}^{(n)} |_{j}^{k} = \frac{k_{\ell}^{(n+1)}}{\ell(n+1)} - \frac{j_{\ell}^{(n+1)}}{\ell(n+1)}.$$
(11)

Theorem 2.7 [6] (*m*-series formula) Let $m \in \mathbb{N}(2)$. Then, we find

$$\sum_{r=m}^{\left\lfloor\frac{k}{\ell}\right\rfloor} \frac{(r-1)^{(m-1)}}{(m-1)!} u(k-r\ell) = \Delta_{\ell}^{-m} u(k) |_{(m-1)\ell+j}^{k} + \sum_{t=1}^{m-1} \sum_{\{m_{1},\dots,m_{t}\} \in \ell(L_{m-1})} (-1)^{t} \times (\Delta_{\ell}^{-m_{1}} u((m_{1}-1)\ell+j)) \frac{k_{\ell}^{(m-m_{t})}}{(m-m_{t})!\ell^{m-m_{t}}} \prod_{i=2}^{t} \frac{((m_{i}-1)\ell+j)_{\ell}^{(m_{i}-m_{i-1})}}{(m_{i}-m_{i-1})!\ell^{m_{i}-m_{i-1}}} |_{(m-1)\ell+j}^{k}$$
(12)

LHS of (12) gives m-series and RHS provides the value of the m-series to u(k).

3. SUMMATION OPERATOR

In this section, we introduce some summation notations for representing m-series in a simple manner. (i) $\mathbb{T}_{oo} = \{(1,0), (0,1)\}, (ii) \quad \mathbb{T}_{oe} = \{(1,0), (0,1), (1,1)\},\$

(iii)
$$\mathbb{T}_{eo} = \{(1,0), (0,1), (1,-1)\}, (iv) \mathbb{T}_{ee} = \{(1,0), (0,1), (1,1), (1,-1)\}, (v) \quad ((n_2)) = {\binom{n_2}{n_2}}^{-uv(\frac{u-v}{u^2+v^2})}, (vi) \quad ((n_3)) = {\binom{n_3}{n_3}}^{uv(\frac{u+v}{u^2+v^2})}$$

(1) If n_2 and n_3 are odd positive integers, then

$$\sum_{(n_2,n_3)}^{s,c} = \frac{(-1)^{\frac{n_2-1}{2}}}{2^{n_2+n_3-1}} \sum_{s_2=0}^{\frac{n_2-1}{2}} \sum_{s_3=0}^{\frac{n_2-1}{2}} (-1)^{s_2} \frac{n_2^{(s_2)}}{s_2!} \frac{n_3^{(s_3)}}{s_3!}$$

(2) If n_2 and n_3 are even positive integers, then

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$$\sum_{[n_2,n_3]}^{s,c} = \frac{(-1)^{\frac{n_2}{2}}}{2^{n_2+n_3-1}} \sum_{s_2=0}^{\frac{n_2-2}{2}} \sum_{s_3=0}^{\frac{n_3-2}{2}} (-1)^{s_2} \frac{n_2^{(s_2)}}{s_2!} \frac{n_3^{(s_3)}}{s_3!}$$

Operators used for product of circular, k-factorial and exponential functions.

$$\sum_{n_1(n_2,n_3)}^{s,c,m+s_1} = \frac{(-1)^{\frac{n_2-1}{2}}}{2^{n_2+n_3-1}} \sum_{s_1=0}^{n_1} \sum_{s_2=0}^{\frac{n_2-1}{2}} \sum_{s_3=0}^{m_3-1} \sum_{s_4=0}^{m+s_1} \frac{n_1^{(s_1)}n_2^{(s_2)}n_3^{(s_3)}(m+s_1)^{(s_4)}}{(-1)^{s_1+s_2+s_4}s_1!s_2!s_3!s_4!}$$

Operators used for product of circular functions with exponential alone.

$$\sum_{(n_2,n_3)m}^{s,c,m} = \frac{(-1)^{\frac{n_2-1}{2}}}{2^{n_2+n_3-2}} \sum_{s_1=0}^{m} \sum_{s_2=0}^{\frac{n_2-1}{2}} \sum_{s_3=0}^{n_3-1} (-1)^{s_1+s_2} \frac{m^{(s_1)}}{s_1!} \frac{n_2^{(s_2)}}{s_2!} \frac{n_3^{(s_3)}}{s_3!}$$

In this section we assume that $P\ell_i$, $\overline{P}\ell_i$, $\left(\frac{P+\overline{P}}{2}\right)\ell_i$, $\left(\frac{P-\overline{P}}{2}\right)\ell_i$ are not multiple of 2π for

 $i = 1, 2, \dots, n$, $m_{(r)} = m(m+1)(m+2)\cdots(m+(r-1))$ and *s*, *p* and *q* are any real numbers. We find m-series of product of polynomial factorial, sine and cosine functions.

4.1 Inverse operator on product of two functions

Theorem 4.1 If n_2 and n_3 are odd positive integers, then we have $\Delta_{\ell}^{-m}(k_{\ell}^{(n_1)}\sin^{n_2}pk\cos^{n_3}qk)$

$$=\sum_{n_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}^{s,c,m+s_{1}}\sum_{n_{0}(u,v)\in\mathbb{T}_{oo}}m_{(s_{1})}\ell^{s_{1}}k_{\ell}^{(n_{1}-s_{1})}\frac{\sin(\frac{uP+vP}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(1-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}}.$$
(13)

Proof. Take $f_1(k) = k_{\ell}^{(1)} \sin^{n_2} p k \cos^{n_3} q k$. Using (7), (8), (9) and changing the powers of sin and cos into linear, we find that

$$\Delta_{\ell}^{-1}f_{1}(k) = \sum_{1(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}^{s,c,1+s_{1}} \sum_{1(s_{1})\in\mathbb{T}_{oo}} 1_{(s_{1})}\ell^{s_{1}}k_{\ell}^{(1-s_{1})} \frac{\sin(\frac{uP+vP}{u^{2}+v^{2}})(k-(1-s_{4})\ell)}{(2(1-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{1+s_{1}}}$$

Applying Δ_{ℓ}^{-1} on both sides to above equation, we get

$$\Delta_{\ell}^{-2} f_1(k) = \sum_{1(n_2, n_3)(u, v) \in \mathbb{T}_{oo}}^{s, c, 2+s_1} \sum_{2(s_1)} \ell^{s_1} k_{\ell}^{(1-s_1)} \frac{\sin(\frac{uP + v\overline{P}}{u^2 + v^2})(k - (2 - s_4)\ell)}{(2(1 - \cos(\frac{uP + v\overline{P}}{u^2 + v^2})\ell))^{2+s_1}}$$

Proceeding like this, we arrive

$$\Delta_{\ell}^{-m} f_{1}(k) = \sum_{1(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}^{s,c,m+s_{1}} \sum_{m(s_{1})} \frac{\sin(\frac{uP+vP}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(1-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}}$$

Similarly, if we take $f_2(k) = k_\ell^{(2)} \sin^{n_2} p k \cos^{n_3} q k$, we find

$$\Delta_{\ell}^{-m} f_{2}(k) = \sum_{2(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}^{s,c,m+s_{1}} \sum_{0,o,v} m_{(s_{1})} \ell^{s_{1}} k_{\ell}^{(2-s_{1})} \frac{\sin(\frac{uP+vP}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(1-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}}.$$

Continuing this process (m-inverse) up to n_1 , we get $\Delta_{\ell}^{-m} f_{n_1}(k)$ given in (13).

Theorem 4.2 If n_2 and n_3 are even positive integers, then

$$\Delta_{\ell}^{-m}(k_{\ell}^{(n_{1})}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{n_{1}[n_{2},n_{3}](u,v)\in\mathbb{T}_{ee}}^{s,c}(n_{2})(n_{3})m_{(s_{1})}\ell^{s_{1}}k_{\ell}^{(n_{1}-s_{1})}$$

$$\times\{\frac{\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(1-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}} + n_{1}!\frac{n_{2}^{(n_{2})}}{2}\cdot\frac{n_{3}^{(n_{2})}}{2}\cdot\frac{n_{3}^{(n_{2})}}{2}\cdot\frac{k_{\ell}^{(m+n_{1})}}{(m+n_{1})!\ell^{m+n_{1}}}\}.$$
(14)

Proof. The proof is similar to the proof of Theorem 4.1.

Remark 4.3 The m-series to $u(k) = k_{\ell}^{(n_1)} \sin^{n_2} p k \cos^{n_3} q k$ can be obtained by substituting (14) in Theorem 2.7.

Theorem 4.4 If n_2 and n_3 are odd positive integers, then we have $\Delta_{\ell}^{-m}(e^{-sk}\sin^{n_2}pk\cos^{n_3}qk)$

$$=\sum_{(n_2,n_3)(u,v)\in\mathbb{T}_{oo}}^{s,c,m} \sum_{\ell=0}^{s,c,m} e^{-sk} e^{s_1s\ell} \frac{\sin(\frac{uP+vP}{u^2+v^2})(k-(m-s_1)\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell))^m}.$$
(15)

Proof. After changing the powers of sin and cos into linear, we get

$$\Delta_{\ell}^{-1}(e^{-sk}\sin^{n_2}pk\cos^{n_3}qk) = \sum_{(n_2,n_3)}^{s,c} \Delta_{\ell}^{-1}(e^{-sk}(\sin Pk + \sin \overline{P}k)) = \text{Impartof} \sum_{(n_2,n_3)}^{s,c} \times \Delta_{\ell}^{-1}(e^{-sk}(e^{iPk} + e^{i\overline{P}k})) = \text{Impartof} \sum_{(n_2,n_3)}^{s,c} (\frac{e^{(iP-s)k}}{e^{(iP-s)\ell} - 1} + \frac{e^{(i\overline{P}-s)k}}{e^{(i\overline{P}-s)\ell} - 1}),$$

After simplification, we get $\Delta_{\ell}^{-1}(e^{-sk}\sin^{n_2}pk\cos^{n_3}qk)$

$$=\sum_{(n_2,n_3)(u,v)\in\mathbb{T}_{oo}}^{s,c,1} \sum_{e^{-sk}} e^{-sk} e^{s_1s\ell} \frac{\sin(\frac{uP+vP}{u^2+v^2})(k-(1-s_1)\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell))}$$
(16)

Applying Δ_{ℓ}^{-1} on both sides to equation (16), we get

$$\Delta_{\ell}^{-2}(e^{-sk}\sin^{n_2}pk\cos^{n_3}qk)$$

$$=\sum_{(n_2,n_3)(u,v)\in\mathbb{T}_{oo}}^{s,c,2} \sum_{e^{-sk}} e^{s_1s\ell} \frac{\sin(\frac{uP+vP}{u^2+v^2})(k-(2-s_1)\ell)}{(2(\cosh s\ell-\cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell))^2}$$

Continuing this process upto m-inverse, we get (15).

Remark 4.5 The m-series to $u(k) = e^{-sk} \sin^{n_2} p k \cos^{n_3} q k$ can be obtained by substituting (15) in Theorem 2.7.

4.2 Inverse operator on product of three functions

Theorem 4.6 If n_2 and n_3 are odd positive integers, then

$$\Delta_{\ell}^{-m}(k_{\ell}^{(n_{1})}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{n_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}\sum_{\ell}\frac{k_{\ell}^{(n_{1}-s_{1})}m_{(s_{1})}}{\ell^{-s_{1}}e^{s(k+s_{1}\ell)}}\frac{e^{s_{4}s\ell}\sin(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(\cosh s\ell-2\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}}.$$

$$(17)$$
Proof. $\Delta_{\ell}^{-1}(k_{\ell}^{(1)}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{(n_{2},n_{3})}\sum_{\ell}\Delta_{\ell}^{-1}(k_{\ell}^{(1)}e^{-sk}(\sin Pk+\sin\overline{P}k)) = \sum_{1(n_{2},n_{3})}\sum_{(u,v)\in\mathbb{T}_{oo}}1_{(s_{1})}\ell^{s_{1}}k_{\ell}^{(1-s_{1})}e^{-s(k+s_{1}\ell)}e^{s_{4}s\ell}\frac{\sin(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(1-s_{4})\ell)}{(2(\cosh s\ell-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{1+s_{1}}}$
Applying Δ_{ℓ}^{-1} on both sides, we get $\Delta_{\ell}^{-2}(k_{\ell}^{(1)}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{1(n_{2},n_{3})}\sum_{(u,v)\in\mathbb{T}_{oo}}2_{(s_{1})}\ell^{s_{1}}k_{\ell}^{(1-s_{1})}e^{-s(k+s_{1}\ell)}e^{s_{4}s\ell}\frac{\sin(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(2-s_{4})\ell)}{(2(\cosh s\ell-\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(2-s_{4})\ell)}$

Then continuing this process, we get $\Delta_{\ell}^{-m}(k_{\ell}^{(1)}e^{-sk}\sin^{n_2}pk\cos^{n_3}qk) = \sum_{1(n_2,n_3)}^{s,c,m+s_1}$

$$\times \sum_{(u,v)\in\mathbb{T}_{oo}} m_{(s_1)} \ell^{s_1} k_{\ell}^{(1-s_1)} e^{-s(k+s_1\ell)} e^{s_4s\ell} \frac{\sin(\frac{uP+vP}{u^2+v^2})(k-(m-s_4)\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell))^{m+s_1}}$$

Similarly we can obtain, $\Delta_{\ell}^{-m}(k_{\ell}^{(2)}e^{-sk}\sin^{n_2}pk\cos^{n_3}qk) = \sum_{2(n_2,n_3)}^{s,c,m+s_1}$

$$\times \sum_{(u,v)\in\mathbb{T}_{oo}} m_{(s_1)} \ell^{s_1} k_{\ell}^{(2-s_1)} e^{-s(k+s_1\ell)} e^{s_4s\ell} \frac{\sin(\frac{uP+v\overline{P}}{u^2+v^2})(k-(m-s_4)\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell))^{m+s_1}}$$

Continuing this process up to m-inverse for n_1 , we get equation (17).

Theorem 4.7 If n_2 is an odd and n_3 is an even positive integers, then

$$\Delta_{\ell}^{-m}(k_{\ell}^{(n_1)}e^{-sk}\sin^{n_2}pk\cos^{n_3}qk) = \sum_{n_1(n_2,n_3)(u,v)\in\mathbb{T}_{oe}}^{s,c,m+s_1}(n_2)((n_3))m_{(s_1)}$$

$$\times \frac{k_{\ell}^{(n_{1}-s_{1})}e^{s_{4}s_{\ell}}}{\ell^{-s_{1}}e^{s(\ell+s_{1}^{\ell})}} \frac{\sin(\frac{uP+vP}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}}.$$
(18)

Similarly, we can obtain equation for n_2 even and n_3 odd positive integers.

Theorem 4.8 If n_2 and n_3 are even positive integers, then

$$\Delta_{\ell}^{-m}(k_{\ell}^{(n_{1})}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{n_{1}[n_{2},n_{3}](u,v)\in\mathbb{T}_{ee}}^{s,c,m+s_{1}}\sum_{(n_{2})\in\mathbb{T}_{ee}}((n_{2}))((n_{3}))\frac{k_{\ell}^{(n_{1}-s_{1})}m_{(s_{1})}}{\ell^{-s_{1}}e^{s(k+s_{1}\ell)}}$$

$$\times(\frac{e^{s_{4}s\ell}\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2\cosh s\ell-2\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell)^{m+s_{1}}} + \frac{n_{2}^{(n_{2})}}{2!}\frac{n_{3}^{(n_{2})}}{2!}\frac{n_{3}^{(n_{2})}}{2!}\frac{1}{2(e^{-s\ell}-1)^{m+s_{1}}}).$$
(19)

Corollary 4.9 If n_2 and n_3 are odd positive integers, then

$$\Delta_{\ell}^{-m}(k^{n_{1}}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{r_{1}=1}^{n_{1}}\sum_{r_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}} \frac{S_{r_{1}}^{n_{1}}k_{\ell}^{(r_{1}-s_{1})}m_{(s_{1})}}{\ell^{r_{1}-n_{1}}\ell^{-s_{1}}e^{s(k+s_{1}\ell)}} \frac{e^{s_{4}s\ell}\sin(UV)(k-(m-s_{4})\ell)}{(2\cosh s\ell - 2\cos(UV)\ell)^{m+s_{1}}}.$$
(20)

Proof. The proof follows by applying (ii) of (4) in equation 17.

Remark 4.10 Hereafter we denote $\Pi(t) = \prod_{i=2}^{t} \frac{((m_i - 1)\ell + j)_{\ell}^{(m_i - m_{i-1})}}{(m_i - m_{i-1})!\ell^{m_i - m_{i-1}}}.$

Theorem 4.11 If n_1 and n_2 are odd positive integers, then the m-series to (17) is

$$\sum_{r=m}^{\lfloor \frac{k}{\ell} \rfloor} \frac{(r-1)^{(m-1)}}{(m-1)!} (k-r\ell)_{\ell}^{(n_{1})} e^{-s(k-r\ell)} \sin^{n_{2}} p(k-r\ell) \cos^{n_{3}} q(k-r\ell) = \sum_{n_{1}(n_{2},n_{3})}^{s,c,m+s_{1}} \sum_{(u,v)\in\mathbb{T}_{oo}} \frac{m_{(s_{1})} k_{\ell}^{(n_{1}-s_{1})}}{\ell^{-s_{1}} e^{s(k+s_{1}\ell)}} e^{s_{4}s\ell} \frac{\sin(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(m-s_{4})\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m+s_{1}}} |_{(m-1)\ell+j}^{k} + \sum_{t=1}^{m-1} \sum_{t=1}^{s} \sum_{\{m_{t}\}\in t(L_{m-1})n_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}} (-1)^{t} (m_{1})_{(s_{1})} \frac{((m_{1}-1)\ell+j)_{\ell}^{(n_{1}-s_{1})}\ell^{s_{1}}e^{s_{4}s\ell}}{e^{s((m_{1}-1)\ell+j)+s_{1}\ell)}} \\ \times \frac{\sin(\frac{uP+v\overline{P}}{u^{2}+v^{2}})((s_{4}-1)\ell+j)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{m_{1}+s_{1}}} \frac{\Pi(t)k_{\ell}^{(m-m_{t})}}{(m-m_{t})!\ell^{m-m_{t}}}|_{(m-1)\ell+j}^{k}.$$
(21)

Proof. The proof is obtained by substituting (17) in Theorem (2.7).

Remark 4.12 when $n_3 = 0$ in (21) we will get $\Delta_{\ell}^{-m} [k_{\ell}^{(n_1)} e^{-sk} \sin^{n_2} pk]$ and when $n_2 = 0$ in (21) we will get $\Delta_{\ell}^{-m} [k_{\ell}^{(n_1)} e^{-sk} \cos^{n_3} pk]$.

The following example illustrates a 2-series to $k_{\ell}^{(3)}e^{-2k}\sin^45k\cos^43k$.

Example 4.13 Consider the case m = 2, p = 6, q = 5, s = 2, $n_1 = 3$, $n_2 = 4$ and $n_3 = 4$, let $P = (6(3-2r_1)+5(3-2r_2))$ and $\overline{P} = (6(3-2r_1)-5(3-2r_2))$ In this case, $L_2 = \{1,2\}, 1(L_2) = \{\{1\},\{2\}\}, and$

LHS =
$$\sum_{r=2}^{l-1} \frac{(r-1)^{(2-1)}}{(2-1)!} (k-r\ell)^{(3)}_{\ell} e^{-s(k-r\ell)} \sin^4 p(k-r\ell) \cos^4 q(k-r\ell)$$

RHS is the sum of the terms, (i) $\Delta_{\ell}^{-2} (k_{\ell}^{(3)} e^{-2k} \sin^4 p k \cos^4 q k) \Big|_{\ell+j}^k$ and

(ii) $\Delta_{\ell}^{-1}(j_{\ell}^{(3)}e^{-2j}\sin^4 pj\cos^4 qj) \times \frac{k_{\ell}^{(1)}}{\ell}\Big|_{\ell+j}^k$, where

$$\Delta_{\ell}^{-2}(k_{\ell}^{(3)}e^{-sk}\sin^{n_{2}}pk\cos^{n_{3}}qk) = \sum_{3[n_{2},n_{3}](u,v)\in\mathbb{T}_{ee}}^{s,c,2+s_{1}}\sum_{(n_{2},n_{3}](u,v)\in\mathbb{T}_{ee}}(n_{2}))((n_{3}))\frac{k_{\ell}^{(3-s_{1})}2_{(s_{1})}}{e^{s(k+s_{1}\ell)}\ell^{-s_{1}}}$$
$$\times(\frac{e^{s_{4}s\ell}\cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})(k-(2-s_{4})\ell)}{(2(\cosh s\ell - \cos(\frac{uP+v\overline{P}}{u^{2}+v^{2}})\ell))^{2+s_{1}}} + \frac{n_{2}^{(n_{2})}}{2!}\frac{n_{3}^{(n_{2})}}{2!}\frac{1}{2!}\frac{1}{2(e^{-s\ell}-1)^{2+s_{1}}}).$$

5. LAPLACE TUNING TRANSFORM AND ITS APPLICATIONS

In this section we derive expression for Laplace Tuning Transform for the functions (input signals) sine, cosine and polynomial factorials, and discuss its applications in DSP.

Theorem 5.1 If n_2 and n_3 are odd positive integers, then $(i)L_{\ell}(\sin^{n_2}pk\cos^{n_3}qk)$

$$=\ell\sum_{r=1}^{\infty}(e^{-sr\ell}\sin^{n_2}pr\ell\cos^{n_3}qr\ell)=\sum_{(n_2,n_3)(u,v)\in\mathbb{T}_{oo}}^{s,c,1}\frac{-\ell\sin(\frac{uP+vP}{u^2+v^2})(s_1-1)\ell}{2(\cosh s\ell-\cos(\frac{uP+v\overline{P}}{u^2+v^2})\ell)}$$

(ii)_k $L_{\ell}(\sin^{n_2}pk\cos^{n_3}qk) = -\ell \times \sum_{r=0}^{\infty} (e^{-s(k+r\ell)}\sin^{n_2}p(k+r\ell)\cos^{n_3}q(k+r\ell))$

$$= -\ell \times \sum_{(n_2, n_3)(u, v) \in \mathbb{T}_{oo}}^{s, c, 1} \sum_{e^{-sk}} e^{s_1 s\ell} \frac{\sin(\frac{uP + vP}{u^2 + v^2})(k - (1 - s_1)\ell)}{2(\cosh s\ell - \cos(\frac{uP + v\overline{P}}{u^2 + v^2})\ell)}$$

Proof. The proof of (i) follows by taking m = 1 in (15), multiplying by ℓ and applying limits for k from 0 to ∞ .

Example 5.2 Taking $n_2 = 1, n_3 = 1, p = 2$ and q = 3 in Theorem (5.1), we obtain

$$L_{\ell}(\sin 2k\cos 3k) = \frac{\ell^2 \sin 5\ell}{2(\cosh s\ell - \cos 5\ell)} - \frac{\ell^2 \sin \ell}{2(\cosh s\ell - \cos \ell)} = \ell \sum_{r=0}^{\infty} e^{-sr\ell} \sin 2r\ell \cos 3r\ell$$

which is verified for $\ell = 0.5$ and s = 5 by MATLAB coding given below : syms r

$$symsum(0.5 * exp(-5 * 0.5 * r) * sin(2 * 0.5 * r) * cos(3 * 0.5 * r), r, 0, inf) = (0.25 * sin(5 * 0.5))./(2 * (cosh(5 * 0.5) - cos(5 * 0.5))) - (0.25 * sin(0.5))./(2 * (cosh(5 * 0.5) - cos(0.5))))$$



Theorem 5.3 If n_2 and n_3 are odd positive integers, then

$$(i)L_{\ell}(k^{n_{1}}\sin^{n_{2}}pk\cos^{n_{3}}qk) = -\ell\sum_{r=0}^{\infty}(r\ell)^{n_{1}}e^{-sr\ell}\sin^{n_{2}}pr\ell\cos^{n_{3}}qr\ell$$

$$=\sum_{r_{1}=1}^{n_{1}}\sum_{r_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}\frac{S_{r_{1}}^{n_{1}}\ell^{n_{1}-r_{1}}\mathbf{1}_{(r_{1})}}{\ell^{-r_{1}}e^{sr_{1}\ell}}\frac{e^{s_{4}s\ell}\sin(UV)(s_{4}-1)\ell}{(2\cosh s\ell - 2\cos(UV)\ell)^{1+r_{1}}}.$$

$$(ii)_{k}L_{\ell}(k^{n_{1}}\sin^{n_{2}}pk\cos^{n_{3}}qk) = -\ell \times \sum_{r=0}^{\infty}(k+r\ell)^{n_{1}}e^{-s(k+r\ell)}\sin^{n_{2}}p(k+r\ell)\cos^{n_{3}}q(k+r\ell)$$

$$=\sum_{r_{1}=1}^{n_{1}}\sum_{r_{1}(n_{2},n_{3})(u,v)\in\mathbb{T}_{oo}}\frac{S_{r_{1}}^{n_{1}}k_{\ell}^{(r_{1}-s_{1})}\mathbf{1}_{(s_{1})}}{\ell^{r_{1}-n_{1}}\ell^{-s_{1}}e^{s(k+s_{1}\ell)}}\frac{e^{s_{4}s\ell}\sin(UV)(k-(1-s_{4})\ell)}{(2\cosh s\ell - 2\cos(UV)\ell)^{1+s_{1}}}.$$

Proof. The proof of (i) follows by taking m = 1 in (20), multiplying by ℓ and applying limits for k from 0 to ∞ .

Example 5.4 Taking $n_1 = 2, n_2 = 1, n_3 = 1, p = 5$ and q = 7 in Theorem (5.3), we obtain

$$L_{\ell}(k^{2}\sin 5k\cos 7k) = (-\ell)\sum_{r=0}^{\infty} (r\ell)^{2} e^{-sr\ell} \sin 5r\ell \cos 7r\ell = \sum_{r_{1}=1}^{2} \sum_{3(1,1)}^{s,c,1+r_{1}} \sum_{(u,v)\in\mathbb{T}_{oo}} \frac{S_{r_{1}}^{2}\ell^{2-r_{1}}\mathbf{1}_{(r_{1})}}{\ell^{-r_{1}}e^{sr_{1}\ell}} \frac{e^{s_{4}s\ell}\sin(UV)(s_{4}-1)\ell}{(2\cosh s\ell - 2\cos(UV)\ell)^{1+r_{1}}}$$

which is verified for $\ell = 0.5$ and s = 10 by MATLAB.





Conclusion: From the outcome of our findings, we observe with the help of the diagrams generated by MATLAB that LTTT gives innumerable outcomes by varying the Time Tuning Factor ℓ for the given input signal and this enables us to make a choice for an optimal one in DSP.

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